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A Two Level Shunt Active Power Filter without PLL for Industrial Loads

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ABSTRACT

Background: To improve the quality of the electrical power to the consumers, many power electronic based solutions are available. In the case of active power factor compensation, grid tied voltage source inverter based Shunt Active Power Filter is one of the versatile solution. But in this system most of the control strategies demand generation of unit vector based on the grid voltage for vector orientation. But presence of harmonics in grid voltages and the level of noise in the grid voltage sensing circuits will make estimation of unit vector difficult and PLL circuits will become a necessity. This paper presents a simple and efficient method to evaluate the unit vector without the use of PLL or any hardware filter. This method was used in d-q reference frame control strategy for shunt active filter and was simulated with a 100 kVAR system. The MATLAB/SIMULINK simulation results are presented to validate the effectiveness of this control strategy.

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INTRODUCTION

Process industries use wide range of variable speed motor drives, ac plants, UPS systems and various power electronic converters to improve system efficiency and hence the productivity. These loads draw reactive power from the grid. Excess reactive power will result in poor voltage regulation and poor utilization of ac network, mainly the distribution transformer for the plant. These constraints make it necessary to compensate reactive power. Traditionally, passive elements like ac capacitor banks are used for reactive power compensation. But the effectiveness of these capacitors is limited. Moreover they have some disadvantages like drawing fixed amount of leading reactive power irrespective of the load requirements. For continuously varying load, which is the case in most of the process industries, these fixed capacitor banks fail to control the power factor effectively.

Thyristor switched or mechanically switched capacitor banks are also used to improve this situation. But they too cannot maintain the power factor near unity when the load varies rapidly. Also, the life span of these passive elements is less due to load harmonics, current sinking and these results in regular maintenance. Due to these limitations of passive filters, various active filters have been reported for reactive power compensation.

In this paper, synchronous d-q reference frame based control strategy is presented for a 100 kVAR two-level shunt active filter. The control law is derived for the compensation of reactive power drawn from grid. This paper also presents a simple and efficient method of unit vector generation, which will minimize the effect of harmonics and noise present in the grid voltage feedback

Shunt Active Power Filter:

The basic block diagram of a three phase shunt active filter is shown in Fig. 1. This STATCOM is primarily a voltage source inverter connected to Point of control, Several control strategies, such as Instantaneous reactive power based hysteresis current controller (Akagi, H., 1984), sliding mode controller (Sateio, S., 1997; Marthinus, G.F., 2004), synchronous reference frame controller (Sensarma, P.S., 2000) etc. have been proposed and developed for a three phase voltage source inverter based shunt compensator. Most of these control strategies require computations using the grid voltage sensed at PCC.

But the source harmonics and the noise in the feedback circuit for grid voltage restrict direct use of the voltage signals in the control algorithm. The harmonics present in the grid voltage can be filtered using hardware band-pass filter (Sensarma, P.S., 2000). But, it is very difficult to design such a filter with precise choice of cut-off frequency that would separate out the fundamental without appreciable phase errors. Some

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work has been reported on the use of PLL to remove the effect of harmonics and noise in the grid voltage. But the design of a high performance PLL is not so easy when various non-idealities like multiple zero crossing in the grid voltage are occurring.

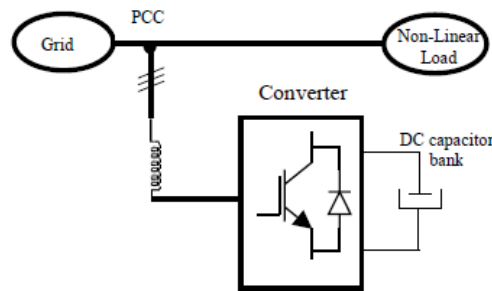


Fig. 1: Grid tied VSI based Shunt Active Power Filter.

Proposed Scheme:

As mentioned earlier, the three phase shunt active filter is a fully controlled three phase boost converter, which is connected to the grid through a three-phase series choke. This converter supplies the required reactive power to the load to maintain the grid side power factor unity. The series choke separates the two voltage sources namely grid and inverter. The capacitor voltage is maintained at a constant value by closed loop control, which regulates the active power drawn from the grid and caters to the internal losses of the system.

Synchronous d-q reference frame based control strategy is proposed to control this converter. This control strategy requires sensing of grid voltage and generation of unit vectors for the orientation along the grid voltages. Fig. 2 explains the method adopted to reduce the effect of harmonics and noise in the grid voltage feedback.

The sensed grid voltages are filtered using a first order digital low pass filter whose corner frequency is ω_c . In the present work, ω_c is chosen to be equal to the nominal grid frequency (50 Hz). So, after filtering, the percentage of h^{th} order harmonics of the sensed grid voltage is reduced by a factor of $\sqrt{\frac{2}{(h^2+1)}}$ (assuming fundamental voltage is always 100%). It is clear that 5th and higher order harmonics, which can normally be present in the line-to-line grid voltage, are reduced considerably using this low pass filter. High frequency noise gets eliminated almost completely. Finally, dividing the filter output signals with their magnitude generates the required unit vectors.

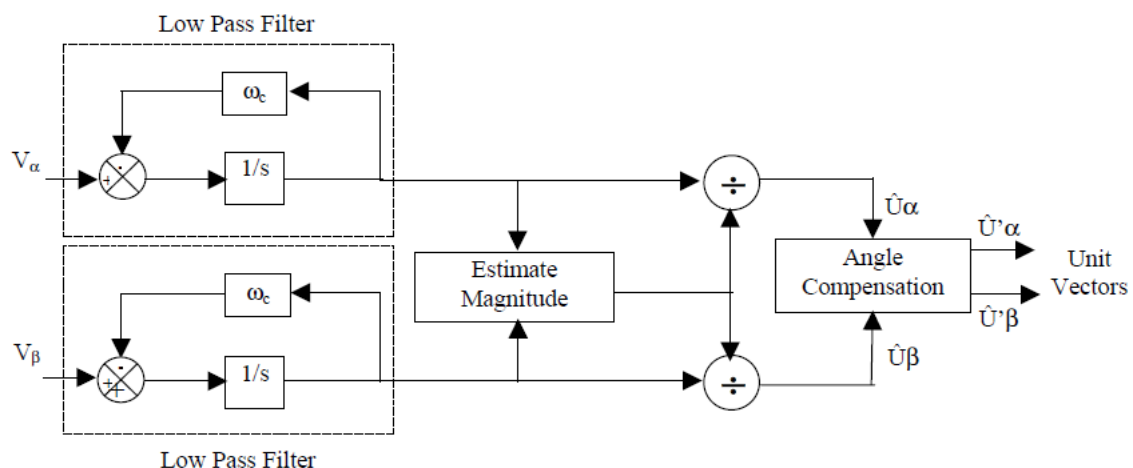


Fig. 2: Block diagram of Unit Vector generation.

However, this low pass filter introduces a phase lag to the unit vector generation. For nominal grid frequency (50Hz), this phase lag is 45° and can be easily compensated (Fig. 2). However, the grid frequency varies with a small range (± 2.5 Hz). So, there is a small phase error in unit vector generation if constant phase compensation is done. In the next section, synchronous d-q frame based control strategy is explained using this unit vector.

Control Strategy:

The control scheme is presented in Fig. 3. The three phase currents of the load and the converter are transformed into the synchronously rotating reference frame using (1) and (2).

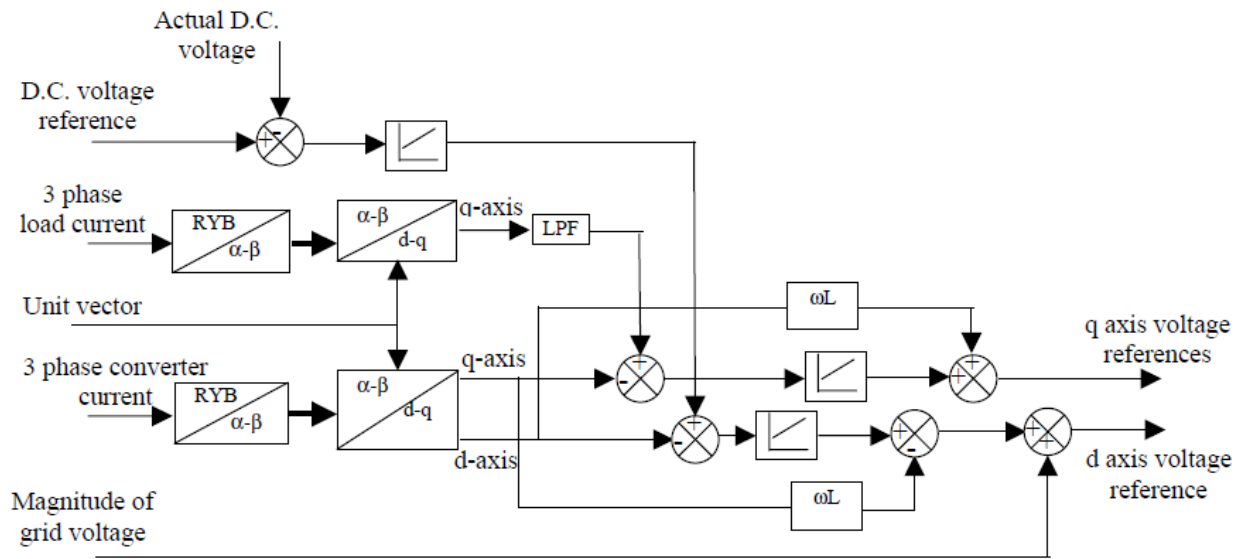


Fig. 3: Control Scheme for Shunt active power filter.

The Unit vector required for the transformation is generated with grid voltages as described earlier. In other words the first step in the control scheme is orienting the converter and grid current along the grid voltage. The load current will be a composite current containing the fundamental and harmonics. After orientation, the fundamental components of d-axis and q-axis currents are the active and reactive parts respectively of the fundamental load currents. For grid reactive power compensation, this fundamental q-axis load current is used as the reference of q-axis current controller of the converter. There is a closed loop PI controller to maintain the dc bus constant. The output of this controller generates the reference of d-axis current controller of the converter.

$$\begin{bmatrix} i_\alpha \\ i_\beta \end{bmatrix} = \begin{bmatrix} 1 & -\frac{1}{2} & -\frac{1}{2} \\ 0 & \frac{\sqrt{3}}{2} & -\frac{\sqrt{3}}{2} \end{bmatrix} \begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix} \quad (1)$$

$$\begin{bmatrix} i_d \\ i_q \end{bmatrix} = \begin{bmatrix} \cos\omega t & -\sin\omega t \\ \sin\omega t & \cos\omega t \end{bmatrix} \begin{bmatrix} i_\alpha \\ i_\beta \end{bmatrix} \quad (2)$$

The control law [6]-[7] along the d and q axis is given below:

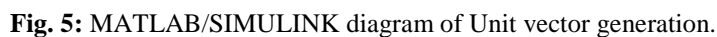
$$V_d = i_d R + L \frac{di_d}{dt} - \omega L i_q + V_g \quad (3)$$

$$V_q = i_q R + L \frac{di_q}{dt} - \omega L i_d \quad (4)$$

Here R and L are the resistance and inductance of the series choke. V_d and V_q are d and q axis voltage commands respectively. There are two PI current controllers to control the d-axis and q-axis current of the converter. The outputs of these current controllers are added with feed forward terms based on (3) and (4) to generate the d-axis and q-axis voltage references for the converter. Finally, d-q voltage references are transformed back to 3-phase stationary voltage references using the unit vectors. These reference signals are fed to the PWM modulator to generate the gate pulses for the converter.

Simulation:

The whole system is simulated in MATLAB. The three-phase source voltages are assumed to be balanced and sinusoidal. A load with highly nonlinear characteristics is considered for the load compensation.



The simulation results at other two phases namely phase B and phase C have also shown similar results as in the figures 7 and 8.

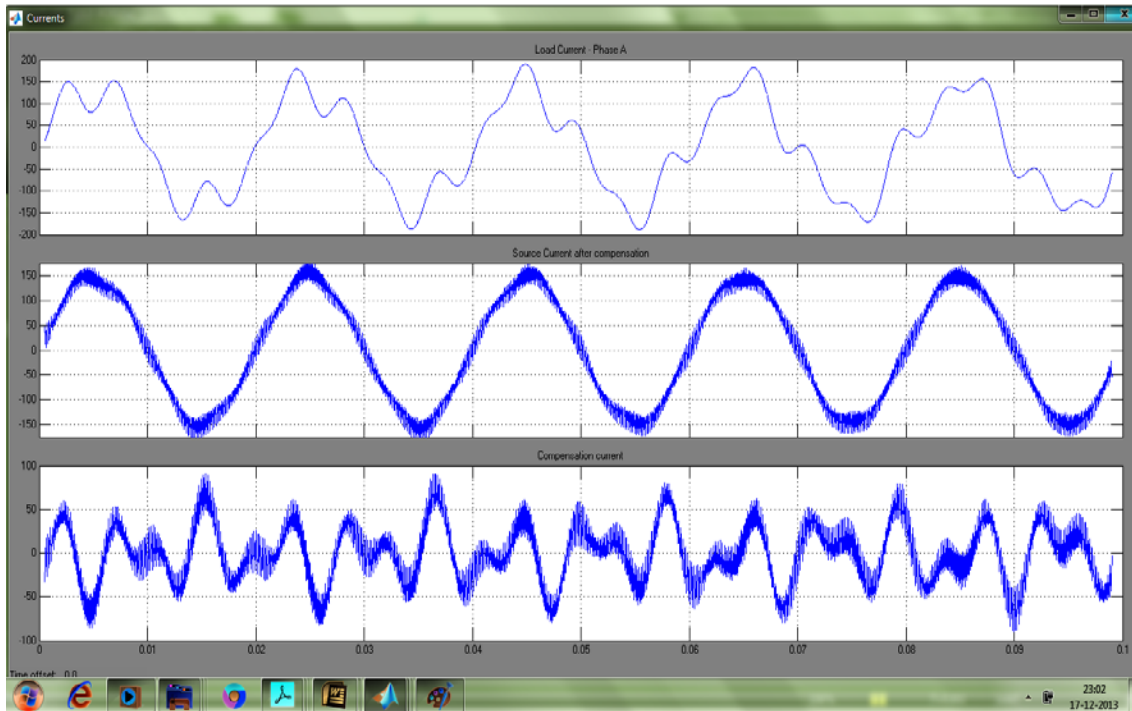


Fig. 6: MATLAB/SIMULINK result of phase 'A' showing the a) load current for 100kVAR load b) source current after compensation c) compensation current injected by the filter into to the point of common coupling.

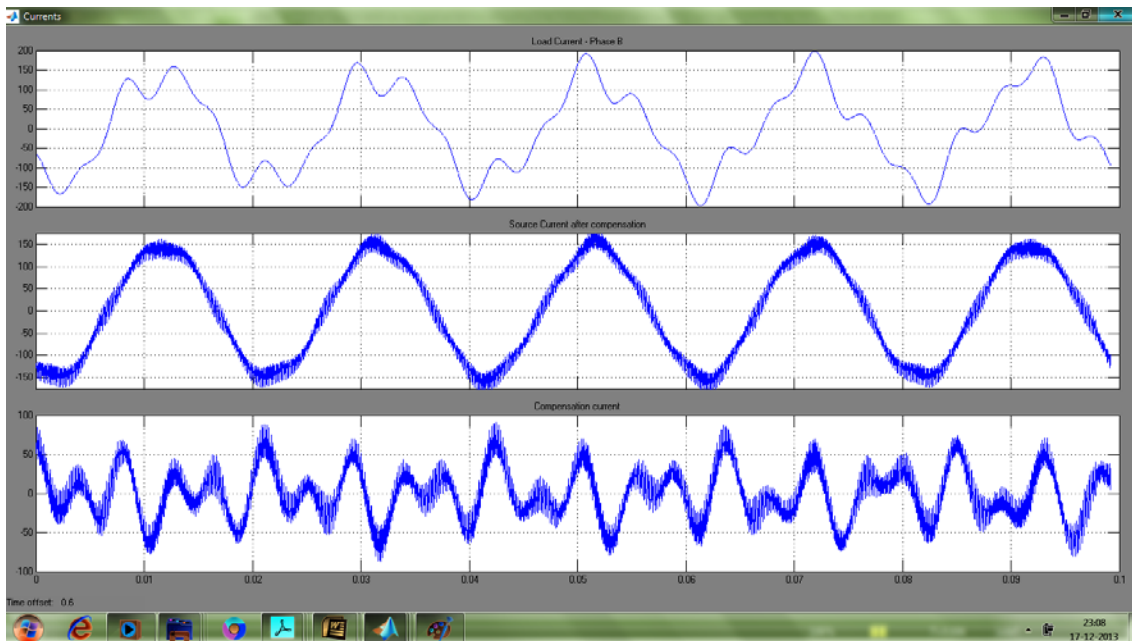


Fig. 7: MATLAB/SIMULINK result of phase 'B' showing the a) load current for 100kVAR load b) source current after compensation c) compensation current injected by the filter into to the point of common coupling.

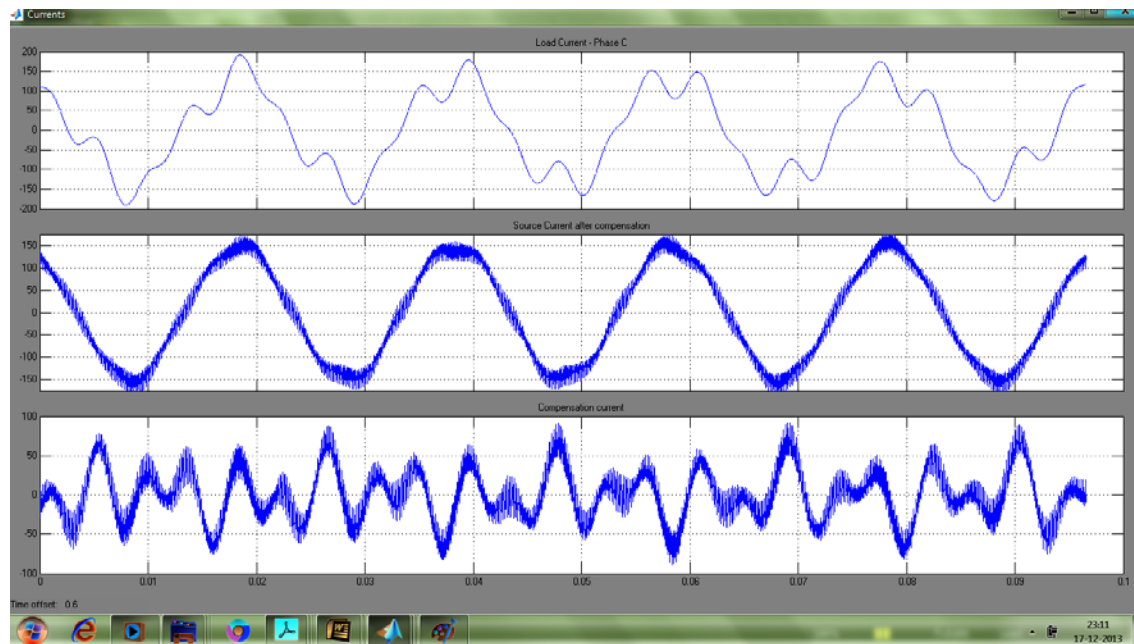


Fig. 8: MATLAB/SIMULINK result of phase 'C' showing the a) load current for 100kVAR load b) source current after compensation c) compensation current injected by the filter into to the point of common coupling.

Conclusion:

This paper proposed an improved algorithm to generate unit vector from the sensed grid voltage with a reduced effect of harmonics and noise present in the feedback signal. This simplified algorithm was used in synchronous d-q reference frame method for the control of shunt active power filter.

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